Electrodynamics II: Lecture 8 EM radiation

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Example: Radiation from antennas







Example: Radiation from antennas



$\vec{\mathbf{E}}$ and $\vec{\mathbf{B}}$ from $\vec{\mathbf{A}}$ and ϕ

Now that we know

$$\phi(\vec{\mathbf{x}},t) = \frac{1}{4\pi\epsilon_0} \int \frac{\rho(\vec{\mathbf{x}}',t_-)}{|\vec{\mathbf{x}}-\vec{\mathbf{x}}'|} d^3 x' , \quad \vec{\mathbf{A}}(\vec{\mathbf{x}},t) = \frac{\mu_0}{4\pi} \int \frac{\vec{\mathbf{J}}(\vec{\mathbf{x}}',t_-)}{|\vec{\mathbf{x}}-\vec{\mathbf{x}}'|} d^3 x'$$
(1)

• We can calculate $\vec{\mathbf{E}} = -\nabla \phi - \partial \vec{\mathbf{A}} / \partial t$ and $\vec{\mathbf{B}} = \nabla \times \vec{\mathbf{A}}$. Note that we'll have to use

$$\nabla \rho(\vec{\mathbf{x}}', t_{-}) = \hat{\mathbf{r}} \frac{d}{dr} \rho = \hat{\mathbf{r}} \left(\frac{\partial \rho}{\partial r} + \frac{\partial \rho}{\partial t_{-}} \frac{dt_{-}}{dr} \right) = \hat{\mathbf{r}} \left(\frac{\partial \rho}{\partial r} + \frac{1}{c} \frac{\partial \rho}{\partial t} \right) \quad (2)$$

where $\vec{\mathbf{r}} = \vec{\mathbf{x}} - \vec{\mathbf{x}}'$.

The electric and magnetic fields turn out to be

$$\vec{\mathbf{E}}(\vec{\mathbf{x}},t) = \frac{1}{4\pi\epsilon_0} \int \left(\frac{\rho(\vec{\mathbf{x}}',t_-)}{r^2} \hat{\mathbf{r}} + \frac{\dot{\rho}(\vec{\mathbf{x}}',t_-)}{cr} \hat{\mathbf{r}} - \frac{\dot{\mathbf{J}}(\mathbf{x}',\mathbf{t}_-)}{c^2r} \right) d^3x'$$
$$\vec{\mathbf{B}}(\vec{\mathbf{x}},t) = \frac{\mu_0}{4\pi} \int \left(\frac{\vec{\mathbf{J}}(\vec{\mathbf{x}}',t_-)}{r^2} \times \hat{\mathbf{r}} + \frac{\dot{\vec{\mathbf{J}}}(\mathbf{x}',\mathbf{t}_-)}{cr} \times \hat{\mathbf{r}} \right) d^3x'$$
(3)

$\vec{E}(\vec{x}, t)$ and $\vec{B}(\vec{x}, t)$: behaviour at large $|\vec{x}|$

- In both, $\vec{\mathbf{E}}(\vec{\mathbf{x}}, t)$ and $\vec{\mathbf{B}}(\vec{\mathbf{x}}, t)$, there are terms that behave as $1/r^2$ and there are terms that behave as 1/r. The former are proportional to the sources, the latter are proportional to the rate of change of sources.
- When the sources are confined to a small region |x
 i | < d, then for |x
 i | >> d, the 1/r terms dominate over the others. These are the "radiative" components of the fields.
- The radiative \vec{E} and \vec{B} fields are then

$$\vec{\mathbf{E}}_{rad}(\vec{\mathbf{x}},t) = \frac{1}{4\pi\epsilon_0} \int \left(\frac{[\dot{\rho}(\vec{\mathbf{x}}')]}{cr}\hat{\mathbf{r}} - \frac{[\dot{\mathbf{J}}(\vec{\mathbf{x}}')]}{c^2r}\right) d^3x' \qquad (4)$$
$$\vec{\mathbf{B}}_{rad}(\vec{\mathbf{x}},t) = \frac{\mu_0}{4\pi} \int \left(\frac{[\dot{\mathbf{J}}(\vec{\mathbf{x}}')] \times \hat{\mathbf{r}}}{cr}\right) d^3x' \qquad (5)$$

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\vec{E}_{rad} and \vec{B}_{rad} in terms of \vec{J} only

Given the continuity equation

$$\nabla' \cdot J(\vec{\mathbf{x}}', t) + \dot{\rho}(\vec{\mathbf{x}}', t) = \mathbf{0} , \qquad (6)$$

it is clear that if $[\vec{J}(\vec{x}')]$ is known everywhere, so is $[\dot{\rho}(\vec{x}')]$, and the radiative \vec{E} and \vec{B} fields can be written in terms of \vec{J} only.

Some algebraic manipulation yields

$$\vec{\mathbf{B}}_{\rm rad}(\vec{\mathbf{x}},t) = \frac{1}{4\pi\epsilon_0 c^3} \int \frac{[\vec{\mathbf{J}}(\vec{\mathbf{x}}')] \times \hat{\mathbf{r}}}{r} d^3 x'$$
(7)

$$\vec{\mathbf{E}}_{\rm rad}(\vec{\mathbf{x}},t) = \frac{1}{4\pi\epsilon_0 c^2} \int \frac{([\vec{\mathbf{J}}(\vec{\mathbf{x}}')] \times \hat{\mathbf{r}}) \times \hat{\mathbf{r}}}{r} d^3 x' \tag{8}$$

Note that $\mu_0/(4\pi) = 1/(4\pi\epsilon c^2)$, one can use any combination.

Frequency components of radiation fields

Fourier components of B_{rad}(x, t) and E_{rad}(x, t) give the radiation fields in terms of their frequency components:

$$B_{\omega}^{rad}(\vec{\mathbf{x}}) = \frac{-i}{4\pi\epsilon c^2} \int \left(\vec{\mathbf{J}}_{\omega}(\vec{\mathbf{x}}') \times \vec{\mathbf{k}}\right) \frac{e^{ikr}}{r} d^3x'$$
(9)

$$E_{\omega}^{rad}(\vec{\mathbf{x}}) = \frac{i}{4\pi\epsilon_0 c} \int \left(\hat{\mathbf{r}} \times (\vec{\mathbf{J}}_{\omega}(\vec{\mathbf{x}}') \times \vec{\mathbf{k}})\right) \frac{e^{ikr}}{r} d^3 x' \quad (10)$$

 If the sources are monochromatic, the above directly give the corresponding **E** and **B** fields. We shall use this in calculating the power radiated by periodically time-varying charges and currents.







Example: Radiation from antennas



Calculation of radiated power

- The Poynting vector is **N** = **E**^{*} × **H**, which gives the power radiated per unit area along it.
- Total energy radiated per unit area normal to N
 is

$$\int_{-\infty}^{\infty} \vec{\mathbf{N}}(t) dt = \int_{-\infty}^{\infty} \vec{\mathbf{E}}^{*}(t) \times \vec{\mathbf{H}}(t) dt \qquad (11)$$
$$= \int_{\omega,\omega',t} \vec{\mathbf{E}}_{\omega}^{\mathrm{rad}*} e^{i\omega t} d\omega \times \vec{\mathbf{H}}_{\omega'} e^{-i\omega' t} d\omega' dt$$
$$= 2\pi \int \vec{\mathbf{E}}_{\omega}^{\mathrm{rad}*} \times \vec{\mathbf{H}}_{\omega}^{\mathrm{rad}} d\omega \qquad (12)$$

• Substituting the expressions for $\vec{\mathbf{E}}_{\omega}^{rad}$ and $\vec{\mathbf{H}}_{\omega}^{rad} = \vec{\mathbf{B}}_{\omega}^{rad}/\mu_0$ obtained earlier, after a bit of algebra, gives

$$\vec{\mathbf{E}}_{\omega}^{\mathrm{rad}*} \times \vec{\mathbf{H}}_{\omega}^{\mathrm{rad}} = \frac{1}{(4\pi)^2} \sqrt{\frac{\mu_0}{\epsilon_0}} \left| \int \left(\vec{\mathbf{J}}_{\omega}(\vec{\mathbf{x}}') \times \vec{\mathbf{k}} \right) \frac{e^{kr}}{r} d^3 x' \right|$$
(13)

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Total energy radiated

Total radiated energy across a surface dS is

$$U = 2\pi \int \vec{\mathbf{E}}_{\omega}^{\mathrm{rad}*} \times \vec{\mathbf{H}}_{\omega}^{\mathrm{rad}} d\omega \cdot d\vec{\mathbf{S}}$$
(14)
$$= \frac{1}{8\pi} \sqrt{\frac{\mu_0}{\epsilon_0}} \int_{\vec{\mathbf{S}},\omega} \left| \int \left(\vec{\mathbf{J}}_{\omega}(\vec{\mathbf{x}}') \times \vec{\mathbf{k}} \right) \frac{e^{ikr}}{r} d^3 x' \right| d\omega \hat{\mathbf{r}} \cdot d\vec{\mathbf{S}}$$
(15)

Taking the surface to be a part of the surface of a sphere of radius *r*, using d**S** = r²∂Ω**r**, one gets

$$\frac{d^2 U}{d\omega d\Omega} = \frac{1}{8\pi} \sqrt{\frac{\mu_0}{\epsilon_0}} \left| \int \left(\vec{\mathbf{J}}_{\omega}(\vec{\mathbf{x}}') \times \vec{\mathbf{k}} \right) e^{ikr} d^3 x' \right|$$
(16)

• For a monochromatic source, one can define average power radiated by averaging over one cycle:

$$\frac{dU}{d\Omega} = \frac{1}{2} \frac{1}{(4\pi)^2} \sqrt{\frac{\mu_0}{\epsilon_0}} \left| \int \left(\vec{\mathbf{J}}_{\omega}(\vec{\mathbf{x}}') \times \vec{\mathbf{k}} \right) e^{ikr} d^3 x' \right|$$
(17)

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The factor of 1/2 comes from averaging.

• When $|\vec{\mathbf{x}}| >> |\vec{\mathbf{x}}'|$, then we have

$$kr = k|\vec{\mathbf{x}} - \vec{\mathbf{x}}'| \approx k|\vec{\mathbf{x}}| - \vec{\mathbf{k}} \cdot \vec{\mathbf{x}}'$$
 (18)

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so that

$$\frac{d^2 U}{d\omega d\Omega} = \frac{1}{8\pi} \sqrt{\frac{\mu_0}{\epsilon_0}} \left| \int \left(\vec{\mathbf{J}}_{\omega}(\vec{\mathbf{x}}') \times \vec{\mathbf{k}} \right) e^{-i\vec{\mathbf{k}}\cdot\vec{\mathbf{x}}'} d^3 x' \right|$$
(19)

This can be used in many instances, for example for radiation from antennas, as will be seen in the next section.









 An antenna may be represented as a monochromatic sinusoidal current with frequency ω:

$$J(x', y', z') = I_0 \delta(y') \delta(z') \frac{\sin[k(L/2 - |x'|)]}{\sin(kL/2)}$$
(20)

An useful result here is

$$\int (\vec{\mathbf{J}} \times \vec{\mathbf{k}}) e^{-i\vec{\mathbf{k}} \cdot \vec{\mathbf{x}}'} d^3 x' \bigg| = \frac{2I_0}{\sin\theta\sin(kL/2)} \left[\cos(kL\cos\theta/2) - \cos(kL/2)\right]$$
(21)

 The radiation from such an antenna can be calculated in closed form in special cases like (i) long wavelength limit: kL ≪ 1 (ii) half-wave antenna: kL = π/2, (iii) full-wave antenna: kL = π.

Recap of topics covered in this lecture

- **Ē** and **B** fields in the presence of moving sources
- Radiative components of **E** and **B**: the 1/r behaviour that dominates at large distances

- Poynting vector and power radiated by EM waves
- Long distance approximation for radiated power
- Antennas as sources of EM radiation